



OBSERVATOR







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Determining radio signals in the presence of noise

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Mass composition study using the radio signals of RD



Works well with <u>inclined air showers</u> ($65^\circ \leq \theta \leq 85^\circ$)

Reconstructing the electromagnetic energy of the shower with RD



1. E-field reconstruction

- Digital to analog conversion, upsampling, Hann window etc.
- Unfolding of the response of the signal-processing chain (LNA, impedance matching, filter amplifiers...)
- Unfolding of the antenna response (NEC-2) to get the E-field (EW, NS, N)

2. Calibrated signals

500 1000

position on $\vec{v} \times \vec{B} / m$

1000 -500

- Decomposition of the E-field in the shower plane coordinate system
- Estimation of signal-to-noise ratio (SNR)

position on $\vec{n} \times \vec{B} / r$

- Estimation of the energy fluence $f [eV m^{-2}]$, the energy deposit per unit area

3. Geomagnetic energy fluence

- Analytic correction of early-late asymmetry
- Parameterized subtraction of charge-excess emission \rightarrow 1-dim LDF

4. Radiation energy

- LDF fit to estimate the geo radiation energy E_{aeo} (energy emitted in form of waves)

- Correction on E_{geo} to compensate for the second-order scaling with the geomagnetic angle and air density at Xmax

5. Elm energy



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Reconstructing the electromagnetic energy of the shower with RD (in a nutshell)



energy deposit per unit area **Energy fluence** *f* [eV m⁻²]



Radio background at Auger



How do we determine the signal (fluence) from the noisy radio measurements?

Current signal and signal uncertainty estimation (Offline)



Estimation of the energy fluence [eV m⁻²]



The method breaks down at low signal-to-noise ratio

100 ns signal window around the peak



Current signal and signal uncertainty estimation (Offline)

Signal-to-noise ratio cut at station level SNR < 7 (10)

$$SNR = \left(\frac{|A_{tot}^{hilb}|_{max}}{V_{RMS}^{noise}}\right)^2$$

The uncertainties on the reconstructed electromagnetic energy (LDF fitting by $\chi^2_{min.}$) $\sigma_{\rm f}^2 = \sigma_{\rm Gauss}^2 + \sigma_{\rm det}^2 + f_{\rm noise}^2$ (backup) are underestimated

We want to achieve a better estimation of the fluence and its uncertainty exploiting a robust mathematical and statistical background.



CAVEAT: Radio measurements have amplitude and phase!

Our **measurement** can be expressed as: sum of **constant known phasor s** and a **random phasor sum**

Marginal density function for **amplitude** only:

$$p_A(a) = \begin{cases} \frac{a}{\sigma^2} \exp\left(-\frac{a^2 + s^2}{2\sigma^2}\right) I_0\left(\frac{as}{\sigma^2}\right) & a > 0\\ 0 & \text{otherwise} \end{cases}$$

with I_0 modified Bessel function $1^{\mbox{\tiny st}}$ kind of 0-order

Proof and details: Chapter 2.9 from J. W. Goodman, Statistical Optics (2015)





$$p_A(a) = \begin{cases} \frac{a}{\sigma^2} \exp\left(-\frac{a^2 + s^2}{2\sigma^2}\right) I_0\left(\frac{as}{\sigma^2}\right) & a > 0\\ 0 & \text{otherwise} \end{cases}$$

a(f) amplitudes of FFT in the signal window

- 1. Find the Hilbert peak
- 2. Hanning window around the peak
- 3. Clipping in the signal window
- 4. FFT of the clipped trace
- 5. Select frequencies in [30-80] MHz





$$p_A(a) = \begin{cases} \frac{a}{\sigma^2} \exp\left(-\frac{a^2 + s^2}{2\sigma^2}\right) I_0\left(\frac{as}{\sigma^2}\right) & a > 0\\ 0 & \text{otherwise} \end{cases}$$

 σ (f) noise level of each frequency

1. N noise windows all over the trace

2. In each window:

- Hanning, Clipping
- FFT, select frequencies in 30-80 MHz

3. Mean value of each frequency

 $\sigma = \mu / \sqrt{(\pi/2)}$

Rayleigh distribution



By storing in Offline the parameters $a(f), \sigma(f)$ we have access to the likelihood L(s).

Get the estimators of **s(f)** by maximizing L(s) (it can be zero!)

$$\hat{f} = \epsilon_0 \,\mathrm{c} \cdot 2 \,\Delta f \,\sum_{i=0}^{N_{\mathrm{bins}}} \left(\hat{s}_{\mathrm{ML}}(f_i) \cdot \Delta t \right)^2$$

Once we get the fluence likelihood, it can be used in the **LDF fitting procedure** instead of χ^2 minimization \rightarrow uncertainties estimation







Conclusions & Outlook

- The fluence estimation using the Rice distribution method shows a smaller bias than the actual Offline method for low SNR values, both for proton and iron simulations

- At higher SNRs, the bias is comparable

- By storing the Rice parameters we have access to the Likelihood

- LDF fitting using the Rice fluence Likelihood will provide uncertainties

- For backward compatibility, we will store also the uncertainty on the fluence obtained by propagating the ones on the s estimators (work in progress, see **backup**)





All irons - EW polarisation (r < 2)







Cut on timing information

 $|(T_{peak})^{REC} - (T_{peak})^{TRUE}| \le 2ns$

in order to compare the goodness of the methods independently of the PulseFinder

Noise windows and Rayleigh distributed noise assumption



Cut of outliers in the frequency bin



Rayleigh distributed bins?



Reconstructed $E_{\mbox{\tiny em}}$ and uncertainty



Goodness of the LDF fits: too small p-values \rightarrow uncertainties probably underestimated Resolution: how good we reconstruct the energy, std of Eem /EemMC in zenith bins

The relative uncertainties do not match the resolution \rightarrow uncertainties underestimated



Signal uncertainty estimation (Offline)

Uncertainty model derived in C. Glaser's PhD thesis

$$\sigma_{\rm f}^2 = \sigma_{\rm Gauss}^2 + \sigma_{\rm det}^2 + f_{\rm noise}^2$$

$$\sigma_{f_{\text{Gauss}}}^{2} = 4|f|\epsilon_{0}c\Delta t V_{\text{RMS}}^{\text{noise}^{2}} + 2(t_{2}-t_{1})(\epsilon_{0}c)^{2}\Delta t V_{\text{RMS}}^{\text{noise}^{4}}$$

uncertainty due to noise after subtracting it, assumes amplitudes are superposition of signal and white noise Gaussian distributed

 $\sigma_{det} = \sigma_{A}^{detector} \cdot 2 \cdot f$ $f \sim A^{2} \quad 5\% \text{ uncertainty on the amplitudes}$ $\sigma_{A}^{detector} = 0.05 \text{ (antenna variation)}$

5.5.2 Uncertainty of the energy fluence

We estimate the uncertainty of the energy fluence by assuming that the measured electric-field amplitude A(t) is the sum of the cosmic-ray radio pulse S(t) and noise e(t). Furthermore, we assume that the noise e(t) is Gaussian distributed with mean $\mu = 0$ and standard deviation $\sigma = \sigma_e$. The energy fluence of A is then given by the equation

$$f(A) = \epsilon_0 c \Delta t \sum_{t_1}^{t_2} A(t_i)^2 = \epsilon_0 c \Delta t \sum_{t_1}^{t_2} \left[S(t_i) + e(t_i) \right]^2 = \epsilon_0 c \Delta t \sum_{t_1}^{t_2} \left[S(t_i)^2 + 2S(t_i)e(t_i) + e(t_i)^2 \right]$$
(5.16)

and the expectation value of f(A) is

$$\langle f(A) \rangle = \epsilon_0 c \Delta t \sum_{t_1}^{t_2} A(t_i)^2 = \epsilon_0 c \Delta t \sum_{t_1}^{t_2} \left[\langle S(t_i)^2 \rangle + 2 \langle S(t_i)e(t_i) \rangle + \langle e(t_i)^2 \rangle \right]$$

$$= \epsilon_0 c \Delta t \sum_{t_1}^{t_2} \left[\langle S(t_i) \rangle^2 + \underbrace{\operatorname{Var}(S(t_i))}_{=0} + 2 \langle S(t_i) \rangle \underbrace{\langle e(t_i) \rangle}_{=0} \right]$$

$$+ 2 \underbrace{\operatorname{Cov}(S(t_i), e(t_i))}_{=0} + \underbrace{\langle e(t_i) \rangle^2}_{=0} + \underbrace{\operatorname{Var}(e(t_i))}_{\sigma_e^2} \right]$$

$$= \epsilon_0 c \Delta t \sum_{t_1}^{\infty} \left[\langle S(t_i) \rangle^2 + \sigma_e^2 \right] .$$

$$(5.17)$$

Hence, the best estimate of the energy fluence of the radio signal S is indeed

$$f(S) = \epsilon_0 c \Delta t \sum_{t_1}^{t_2} \left[A(t_i)^2 - \sigma_e^2 \right]$$
(5.18)

as defined in Eq. (5.8) where σ_e^2 is also calculated from the electric-field trace in a part where no signal is present. Following a similar calculation we can estimate the uncertainty of f(S) by computing $\sigma_f^2 = Var(f) = \langle f^2 \rangle - \langle f \rangle^2$. After several lines of calculation it follows that

$$\sigma_f^2 = 4f \,\epsilon_0 c \,\Delta t \,\sigma_e^2 + 2 \,(\epsilon_0 c)^2 \,(t_2 - t_1) \,\Delta t \,\sigma_e^4 \,. \tag{5.19}$$

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Bias of the s estimators



What we actually need to know to understand the Likelihood estimator and its bias:

•
$$a/\sigma \gtrsim 3 \rightarrow s_{ML} \approx \sqrt{a^2 - \sigma^2}$$

 $\mathcal{N}(\sqrt{a^2 - \sigma^2}, \sigma)$

C Else: $a - 1.5 \sigma < s_{ML}^2 < a + 1.5 \sigma$

Note that the fluence will be always positive ($f = 0 \text{ eV}/\text{m}^2$ in principle possible!)

Bias of the s estimators



Bias of the s estimators



Signal and signal uncertainty estimation with the Rice distribution

Once we get the fluence likelihood, it can be used in the **LDF fitting procedure** instead of χ^2 minimization \rightarrow uncertainties estimation

For **backward compatibility**, we want to store in Offline the fluence and its error (e.g. get the uncertainties of the estimators of **s**(**f**) and propagate them to the fluence).



$$J(s) = -2\ln\left(\frac{L(s)}{L(s_{\rm ML})}\right) \qquad \text{(backup)}$$
$$J(s) = k^2$$
$$\delta = k\sigma \qquad \mathbf{k=1} \qquad \delta = 1/\sqrt{\frac{1}{2}\frac{\partial^2 J}{\partial s^2}}|_{s=s_{\rm ML}}$$

Gaussian approximation: $\delta \sim \sigma$

In non Gaussian approximation we would need to define asymmetrical errors to have a 68% interval

- Ideally we want to define a 68% interval around the estimator. Looking at the cdf of the normalized Likelihood function we can distinguish two main cases:
 - Gaussian approximation (cdf of the estimator ~ 50%): symmetrical errors



Ideally we want to define a 68% interval around the estimator. Looking at the cdf of the normalized Likelihood function we can distinguish two main cases:

- Gaussian approximation (cdf of the estimator ~ 50%): symmetrical errors $\delta \sim \sigma$
- Non-Gaussian: asymmetrical errors



Ideally we want to define a 68% interval around the estimator. Looking at the cdf of the normalized Likelihood function we can distinguish two main cases:

- Gaussian approximation (cdf of the estimator ~ 50%): symmetrical errors $\delta \sim \sigma$



But, we would like to store 1 single value (symm.) in Offline

• There's no analytical form of the cdf available \rightarrow numerical integration

- discrete n values
- integrals in [i, i+1], with i=0,...,n

• Hessian of the cost function
$$J(s) = -2 \ln \left(\frac{L(s)}{L(s_{\rm ML})}\right) \delta$$

 $J(s) \cong J(\hat{s_{\rm ML}}) + \frac{\partial J}{\partial s}|_{s \neq \hat{s_{\rm ML}}} (s - \hat{s_{\rm ML}}) + \frac{1}{2} \frac{\partial^2 J}{\partial s^2}|_{s = \hat{s_{\rm ML}}} (s - \hat{s_{\rm ML}})^2$





Estimators uncertainty: Hessian vs cdf



The two methods are equivalent (as it should)

Estimators uncertainty: Hessian vs cdf



The two methods diverge

Estimators uncertainty: Hessian vs cdf



• The coverage fluctuate around 68% where the Gaussian approximation is valid. Large bias values mostly correspond to an overestimation of the error.



Correlation of the amplitudes of the signal



Gaussian approximation of L(s)



Rice method: the algorithm(s)

root finder between positive and negative value of dL/ds
 minimizing -L with bounded method between smin, smax
 run one of the previous methods, fit s vs a → in principle faster (less accurate...)



For the three of them, we need to study the solutions space

$$\frac{a^2 e^{\frac{-a^2 - s^2}{2\sigma^2}} I_1\left(\frac{as}{\sigma^2}\right)}{\sigma^4} - \frac{as e^{\frac{-a^2 - s^2}{2\sigma^2}} I_0\left(\frac{as}{\sigma^2}\right)}{\sigma^4}$$
$$a/\sigma \leqslant 1.4 \rightarrow s_{\text{ML}} = 0$$
$$a/\sigma \gtrsim 3 \rightarrow s_{\text{ML}} \approx \sqrt{a^2 - \sigma^2}$$

Rice method: the algorithm(s)

Second derivative to find min/max of the first derivative



$$\frac{a\left(a^{2}sI_{0}\left(\frac{as}{\sigma^{2}}\right)-2as^{2}I_{1}\left(\frac{as}{\sigma^{2}}\right)-a\sigma^{2}I_{1}\left(\frac{as}{\sigma^{2}}\right)+s^{3}I_{0}\left(\frac{as}{\sigma^{2}}\right)-s\sigma^{2}I_{0}\left(\frac{as}{\sigma^{2}}\right)\right)e^{-\frac{a^{2}+s^{2}}{2\sigma^{2}}}}{s\sigma^{6}}$$

$$a-1.5\sigma < s_{\mathrm{ML}}^{2} < a+1.5\sigma$$

$$0$$